



Leakage Inductance

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Overview

- This presentation reviews the concepts of leakage impedances which can be represented in terms of leakage inductances and leakage resistances
- Leakage inductances are known to vary with frequency, but this presentation shows how they also vary with time or pulse width
- Only resistive and inductive effects are considered in the presentation



Magnetic Coupling Review



- Two windings are coupled when some of the magnetic flux produced by currents flowing in either of the windings passes through both windings
- Only part of the flux produced by a current in one winding reaches other windings
- Flux which doesn't pass through both windings is called leakage flux
- Magnetic coupling can be modeled in terms of self and mutual impedances



Self and Mutual Impedance Equations



 Z_{11} = Winding 1 Self Impedance Z_{22} = Winding 2 Self Impedance

 $Z_{12} = Z_{21} =$ Mutual Impedance

 $R_{12} = R_{21} =$ Mutual Resistance

 $L_{12} = L_{21} =$ Mutual Inductance



$$v_1 = Z_{11}i_1 + Z_{12}i_2$$

= $(R_{11} + j\omega L_{11})i_1 + (R_{12} + j\omega L_{12})i_2$

$$v_{2} = Z_{21}i_{1} + Z_{22}i_{2}$$
$$= (R_{21} + j\omega L_{21})i_{1} + (R_{22} + j\omega L_{22})i_{2}$$

Impedance Matrix Equation for N Windings

$$\begin{bmatrix} v_1 \\ v_2 \\ \vdots \\ v_N \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} & \cdots & Z_{1N} \\ Z_{21} & Z_{22} & \cdots & Z_{2N} \\ \vdots & \vdots & \ddots & \vdots \\ Z_{N1} & Z_{N2} & \cdots & Z_{NN} \end{bmatrix} \begin{bmatrix} i_1 \\ i_2 \\ \vdots \\ i_N \end{bmatrix}$$

- A set of coupled windings can be modeled with a matrix equation that relates frequency-domain winding voltages and currents with an impedance matrix
- The values of impedance matrix elements can be obtained through FEA simulations or from measurements
- The impedance matrix values vary with frequency



Basic Definition of Leakage Impedance



$$Z_{leak_mn} = Z_{mm} - \frac{Z_{mn}^2}{Z_{nn}}$$

Winding m Winding n shorted

- Leakage impedance is the impedance measured at one winding when another winding is shorted
- Leakage impedances are a function of self and mutual impedances as shown in the equation above
- Consequently, leakage impedances are a property of a pair of windings and generally can't be split up and assigned to individual windings when there are more than two windings



Inductance and Resistance Measurement Options



 LCR meters and network analyzers typically measure the real and imaginary parts of the impedance and then convert that data into series or parallel representations



Measurement of Series Inductance and Resistance



- The impedance of leakage inductances between closely-coupled windings often behave like this circuit where the Q is relatively low
- The series inductance transitions from 5 μ H (Lhf + Llf) at low frequencies to about 1 μ H (Lhf) at high frequencies due to Rlf
- The parallel resistors represent copper losses
- The core losses are minimal for leakage inductances because one winding is shorted



Pulse Measurement of Inductance



- At the beginning of the pulse, the effective inductance is equal to Lhf
- At the end of the pulse, the effective inductance is about equal to Lhf + Llf



Maxwell 2D Transformer Model





All wires 22 AWG

All windings 38 turns

2 layers 2 mil tape between layers, 0.1mm

Core: ETD49/25/16-3C97

Gap: 3 mil spacer

Bobbin: TDK B66368B1020T001



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FEA Self and Mutual Resistances





FEA Self and Mutual Inductances





FEA Leakage Resistances and Inductances





FEA and Measured Leakage Resistances and Inductances





SPICE Mutual Impedance Transformer Model





Equivalent Circuit Model Closely Matches FEA results



- The highest leakage impedances occur when measuring winding 1 with winding 4 shorted because those windings have the greatest separation
- The equivalent circuit model results closely match the FEA data



Leakage Inductance Pulse Test





Leakage Inductance Simulations



• The range for the effective leakage inductance in the time domain is close to the range in the frequency domain



Pulsing Test Fixture





Leakage Inductance From Pulsed Waveforms





Measured and Extracted Leakage Inductance



Measured leakage Inductance at winding 1 with winding 4 shorted

Leakage Inductance extracted from pulsed waveforms at winding 1 with winding 4 shorted



Diode Reverse Recovery Test Circuit

Winding 1 with winding 4 shorted

(Mutual Resistance model not shown)



PWL(0 -1 3u -1 3.01u 1 5u 1 5.01u -1)

.SUBCKT 1N6631 anode cathode .param IS=300n N=1.5 Tau=60n Y0=1 alpha=5MEG va=0.1 Dpn anode N001 Dj .MODEL Dj D(IS {IS} N {N}) Cdiff N002 diff 1 Vdiff diff 0 0V Ediff N003 0 Value {I(Vdiff)} Hcsense N005 0 VHcsense 1 VHcsense N001 N004 0V G_charge_current anode N004 N003 0 1 E_charge_calculator N002 0 VALUE{Tau*(V(N005)-va*V(N003))} G_base_region_current N004 cathode VALUE {(V(N004) - V(cathode))*(Y0+alpha*V(N002)) } .ENDS



Diode Reverse Recovery Comparison



Mutual Resistance Model Enabled

Mutual Resistance Model Disabled

 The mutual resistance model reduces the effective inductance and increases the peak reverse recovery current



Conclusions

- Leakage inductance for closely-coupled winding pairs decreases with frequency
- The inductance decrease is due to skin and proximity effects
- The effective leakage inductance for pulsed waveforms can be determined by dividing the applied voltage by the time derivative of the current
- Derivatives of measured data are typically noisy and need smoothing
- The effective leakage inductance for short pulses is less than for longer
 pulses
- The currents produced by the reverse recovery of fast diodes connected to transformer outputs depend on the high-frequency leakage inductance values

